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The scale of the problem: recovering images of reionization with Generalized Morphological Component Analysis

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ABSTRACT

The accurate and precise removal of 21-cm foregrounds from Epoch of Reionization (EoR) redshifted 21-cm emission data is essential if we are to gain insight into an unexplored cosmological era. We apply a non-parametric technique, Generalized Morphological Component Analysis (GMCA), to simulated Low Frequency Array (LOFAR)-EoR data and show that it has the ability to clean the foregrounds with high accuracy. We recover the 21-cm 1D, 2D and 3D power spectra with high accuracy across an impressive range of frequencies and scales. We show that GMCA preserves the 21-cm phase information, especially when the smallest spatial scale data is discarded. While it has been shown that LOFAR-EoR image recovery is theoretically possible using image smoothing, we add that wavelet decomposition is an efficient way of recovering 21-cm signal maps to the same or greater order of accuracy with more flexibility. By comparing the GMCA output residual maps (equal to the noise, 21-cm signal and any foreground fitting errors) with the 21-cm maps at one frequency and discarding the smaller wavelet scale information, we find a correlation coefficient of 0.689, compared to 0.588 for the equivalently smoothed image. Considering only the pixels in a central patch covering 50 per cent of the total map area, these coefficients improve to 0.905 and 0.605, respectively, and we conclude that wavelet decomposition is a significantly more powerful method to denoise reconstructed 21-cm maps than smoothing.

Key words: methods: statistical – cosmology: theory – dark ages, reionization, first stars – diffuse radiation.

1 INTRODUCTION

When the first ionizing sources appeared 400 Myr after the big bang, the Universe emerged from the ‘Dark Ages’ and began to be reionized. This Epoch of Reionization (EoR) is on the verge of being directly observed for the first time, with a new generation of radio telescopes beginning to see first light [e.g. Low Frequency Array (LOFAR)]¹ (van Haarlem et al., in preparation), Giant Metrewave

Radio Telescope (GMRT),² Murchison Widefield Array (MWA),³ Precision Array to Probe the Epoch of Reionization (PAPER),⁴ 21 Centimetre Array (21CMA)⁵].

The vast majority of EoR observations will take advantage of the 21-cm spectral line – produced by a spin flip in neutral hydrogen (van de Hulst 1945; Ewen & Purcell 1951; Muller & Oort 1951). This 21-cm radiation can be observed interferometrically at radio wavelengths as a deviation from the brightness temperature of the

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¹ <http://www.lofar.org/>

² <http://gmrt.ncra.tifr.res.in/>

³ <http://www.mwatelescope.org/>

⁴ <http://astro.berkeley.edu/dbacker/eor/>

⁵ <http://21cma.bao.ac.cn/>

cosmic microwave background (CMB; Field 1958, 1959; Madau, Meiksin & Rees 1997; Shaver et al. 1999).

Observationally, the 21-cm signal will be accompanied by systematic effects due to the ionosphere and instrument response, system noise, extragalactic foregrounds and Galactic foregrounds (e.g. Jelić et al. 2008, 2010), the latter of which are orders of magnitude larger than the 21-cm signal we wish to detect. The foregrounds must be accurately and precisely removed from the observed data as any error at this stage has the ability to strongly affect the EoR 21-cm signal. Foreground removal and the implications for 21-cm cosmology have been extensively researched over the past decade (e.g. Di Matteo et al. 2002; Oh & Mack 2003; Di Matteo, Ciardi & Miniati 2004; Morales & Hewitt 2004; Zaldarriaga, Furlanetto & Hernquist 2004; Santos, Cooray & Knox 2005; Bowman, Morales & Hewitt 2006; McQuinn et al. 2006; Wang et al. 2006; Gleser, Nusser & Benson 2008; Jelić et al. 2008; Harker et al. 2009, 2010; Liu, Tegmark & Zaldarriaga 2009a; Liu et al. 2009b; Liu & Tegmark 2011, 2012; Mao 2012; Petrovic & Oh 2011; Chapman et al. 2012; Cho, Lazarian & Timbie 2012). This paper concentrates on the spectral fitting of the foregrounds and assumes that bright sources have been accurately removed, for example via a flux cut (Di Matteo et al. 2004). This is a safe assumption since, as recently shown by Trott, Wayth & Tingay (2012), the residuals from the bright source subtraction are expected to be smaller than the thermal noise contribution and so are not a limiting factor.

In our first paper, Chapman et al. (2012), we introduced a method to successfully remove the foregrounds while making only minimal assumptions. By introducing another similarly successful method, we acknowledge the possibility that different methods will be well suited for the extraction of different information from the data and that there is an advantage in having several foreground cleaning methods to apply the data independently to confirm a statistical detection. In this paper we implement another non-parametric method, the sparsity-based blind source separation (BSS) technique Generalized Morphological Component Analysis (GMCA). GMCA provides a complete basis set for foreground removal as opposed to a polynomial fitting method which is not a complete set unless we describe it with a polynomial of the order of the number of frequencies. Polynomial methods rarely utilize orders larger than five and so most polynomial fitting methods may leave the foregrounds incompletely described compared to GMCA.

In Section 2 we summarize the various foreground removal pipelines which have been introduced in the literature so far and then go on to detail the method that we utilize, GMCA. In Section 3 we introduce our data cube and the methods used to produce it before presenting the statistical results in Section 4. In Section 5 we explore the possibility of recovering images of reionization before we set out our conclusions in Section 6.

2 FOREGROUND REMOVAL TECHNIQUES

The statistical detection of the 21-cm reionization signal depends on an accurate and robust method for removing the foregrounds from the total observed signal. For a brief summary of the recent 21-cm foreground removal literature we ask the reader to refer to section 2 of Chapman et al. (2012).

The majority of the literature involves parametric methods, whereby at some point a certain form for the foregrounds is assumed, for example polynomials (e.g. Santos et al. 2005; Bowman et al. 2006; McQuinn et al. 2006; Wang et al. 2006; Gleser et al. 2008; Jelić et al. 2008; Liu et al. 2009a,b; Petrovic & Oh 2011). In contrast, non-parametric methods do not assume a specific form

for the foregrounds, instead allowing the data to determine the foregrounds using more free parameters. While advantageous for poorly constrained data, results are often not as promising as parametric results, though recent methods have challenged this. Harker et al. (2009, 2010) preferentially considered foreground models with as few inflection points as possible, which when applied to simulated LOFAR-EoR data compared very favourably with parametric methods. Similarly, FASTICA as presented in Chapman et al. (2012) accurately recovered the 21-cm power spectra by considering the statistically independent components of the foregrounds. GMCA (Bobin et al. 2007, 2008a,b, 2012) is another non-parametric method which has a greater flexibility through wavelet choice without the sacrifice of the blind nature of the approach. We will show that GMCA not only recovers the power spectra to high accuracy but also that, using wavelet decomposition, the simulated 21-cm signal maps can be recovered exceedingly well after the foreground removal process.

2.1 The GMCA method

The non-parametric method of removing the foregrounds is effectively a BSS problem. Chapman et al. (2012) utilized a statistical approach to source separation, namely FASTICA. This assumed that the components of the foregrounds were statistically independent and non-Gaussian in order to reconstruct the smooth spectral form of the foregrounds and leave a residual signal from which we could identify the 21-cm emission statistics. This statistical pursuit of independence is only one form that BSS techniques take, the other utilizing morphological diversity and sparsity to separate the sources. Zibulevsky & Pearlmutter (2001) proposed a new method of BSS, where one could find a basis set in which the sources to be found would be sparsely represented, i.e. a basis set where only a few of the coefficients would be non-zero. With the sources being unlikely to have the same few non-zero coefficients one could then use this sparsity to more easily separate the mixture. For example, were the 21-cm signal strong enough to detect directly using this method, we would expect it to be sparse on certain scales given the characteristic size of the ionized bubbles, much in the same way the Sunyaev–Zel’dovich (SZ) effect can be detected with this method when analysing CMB data (Bobin et al. 2008a). These bubble sizes change as a function of redshift so the sparse signal from these would arise as a pattern as a function of wavelength. In comparison, the smooth frequency structure of the foregrounds implies that the few sparse non-zero coefficients describing the foregrounds at the same scales as the 21-cm signal would be unchanging with frequency. Given our noise realizations, the 21-cm signal is far too small for this technique to pick it out as a source in its own right. Instead, it is how the foregrounds can be described as different sparse components which enable us to obtain the 21-cm signal and noise as a residual.

The idea of exploiting the sparseness of sources in different bases has evolved into a full and diverse field of applications. The method has evolved to allow sources to have different morphologies, exploit multichannel data and consider different bases for different sources in order to achieve the most sparse representations.

Consider an observation of m maps each constituting t pixels across m channels of observation. The problem to be solved can be stated in the following manner :

$$\mathbf{X} = \mathbf{AS} + \mathbf{N}, \quad (1)$$

where \mathbf{X} is the $m \times t$ matrix representing the observed data, n is the number of sources to be estimated, \mathbf{S} is the signal $n \times t$ matrix to

be determined, \mathbf{A} is the $m \times n$ mixing matrix and \mathbf{N} is the $m \times t$ noise matrix.

As this is a BSS problem, we need to estimate both \mathbf{S} and \mathbf{A} . We seek to find the 21-cm signal as a residual in the separation process, therefore \mathbf{S} represents the foreground signal and, due to the extremely low signal-to-noise ratio of this problem, the 21-cm signal is numerically ignored by the method and can be thought of as an insignificant part of the noise.

We can expand the sources, $\mathbf{S} = \sum_{j=0}^n s_j$, in a wavelet basis which can mathematically be thought of as a matrix of T wavelet waveforms, $\Phi = [\phi_1, \dots, \phi_T]$, such that $\forall j \in \{1, \dots, n\} s_j = \sum_{k=1}^T \alpha_j[k] \phi_k$, s_j is defined to be sparse if only a few of the $\alpha_j[k]$ are significantly non-zero.

The objective of GMCA is to seek an unmixing scheme, through the estimation of \mathbf{A} , which yields the sparsest sources \mathbf{S} in the wavelet domain. This is expressed by the following optimization problem written in the augmented Lagrangian form:

$$\min \left(\frac{1}{2} \|\mathbf{X} - \mathbf{A}\mathbf{A}\Phi\|_F^2 + \lambda \sum_{j=1}^n \|\alpha_j\|_p \right), \quad (2)$$

where typically $\|\alpha\|_p = (\sum_k |\alpha[k]|^p)^{1/p}$; sparsity is generally enforced for $p = 0$ which measures the number of non-zero entries of α (or its relaxed convex version with $p = 1$) and $\|\mathbf{X}\|_F = (\text{trace}(\mathbf{X}^T \mathbf{X}))^{1/2}$ is the Frobenius norm. The problem in equation (2) is solved in an iterative two-step algorithm such that at each iteration q :

(i) Estimation of the \mathbf{S} for \mathbf{A} fixed to $\mathbf{A}^{(q-1)}$

Solving the problem in equation (2) for $p = 0$ assuming \mathbf{A} is fixed to $\mathbf{A}^{(q-1)}$, the sources are estimated as follows:

$$\mathbf{S}^{(q)} = \Delta_\lambda \left(\mathbf{A}^{(q-1)+} \mathbf{X} \Phi^T \right) \Phi,$$

where Δ_λ stands for the hard-thresholding operator with threshold λ ; this operator puts to zero all coefficients with amplitudes lower than λ . In practice, the threshold λ is set to be equal to three times the standard deviation of the noise level to exclude noise coefficients. The term $\mathbf{A}^{(q-1)+}$ denotes the Moore pseudo-inverse of the matrix $\mathbf{A}^{(q-1)}$.

(ii) Estimation of the \mathbf{A} for \mathbf{S} fixed to $\mathbf{S}^{(q)}$

Updating the mixing matrix assuming that the sources are known and fixed to $\mathbf{S}^{(q)}$ is as follows:

$$\mathbf{A}^{(q)} = \mathbf{X} \mathbf{S}^{(q)+}.$$

For more technical details about GMCA, we refer the interested reader to Bobin et al. (2007, 2008a,b, 2012), where it is shown that sparsity, as used in GMCA, allows for a more precise estimation of the mixing matrix \mathbf{A} and more robustness to noise than independent component analysis (ICA)-based techniques.

GMCA provides an efficient method of separating the foreground signal from the noise and 21-cm signal by locating the most sparse components that the foreground signal could be made of in the wavelet basis Φ . From a Bayesian point of view, using this sparsity method is equivalent to having an in-built prior in the model that the foregrounds are sparse over the basis chosen. While a Bayesian evidence analysis could be possible in order to compare different methods as well as different bases, and quantify the overfitting due to more free parameters, we consider it outside the scope of this project.

2.2 Wavelets

The set of basis functions, Φ , used by GMCA comprises wavelet functions.

The Fourier transform is a well-known method of analysing data at different scales with a single set of basis functions – sines and cosines. In reality this confined basis set can obscure information, and so we instead consider an infinite set of basis functions localized in space – the wavelet functions. There are many types of wavelets – with some more localized in space, some smoother and some with fractal structures.

The most common form of wavelet used in astrophysics is the isotropic undecimated wavelet transform (IUWT) which we describe briefly below in reference to the more complete descriptions in literature (e.g. Starck, Murtagh & Bijaoui 1998; Starck & Murtagh 2006). Consider an image with $p \times p$ pixels (where using the previous section's notation $p \times p = t$ and we will refer to the pixel coordinates as $[k, l]$).

We can decompose an image at a particular frequency, c_0 , into a coarse version of itself, c_J , along with a superposition of the original image at different wavelet scales:

$$c_0[k, l] = c_J[k, l] + \sum_{j=1}^J w_j[k, l], \quad (3)$$

where the wavelet coefficient w_j represents the data at scale 2^{-j} .

The decomposition is typically achieved using low-pass 1D filters, which we call h_{1D} , implemented by the ‘à trous’ algorithm:

$$c_{j+1}[k, l] = \sum_q \sum_p h_{1D}[p] h_{1D}[q] c_j[k + 2^j q, l + 2^j p], \quad (4)$$

$$w_{j+1} = c_j[k, l] - c_{j+1}[k, l]. \quad (5)$$

When c_0 can be described by only a few significantly non-zero w_j , we say that c_0 is sparse in that basis.

In this paper we utilize wavelets twice. The first time is within the GMCA algorithm, where we have a choice of different wavelet types that can be used as the basis set. GMCA uses wavelet decomposition to identify the sources, \mathbf{S} , but then returns a data cube with all scales present. In our analysis in Section 5 we wish to look at these results on different scales and so we utilize wavelet decomposition to do this. We will use the IUWT both within GMCA and later to analyse the images at different scales, though we briefly consider other wavelets when we question how much our results depend on this choice of wavelet.

For our GMCA implementation we set the number of decomposition scales to be eight, the maximum allowed by the dimensions of our data cube.

3 SIMULATED EOR DATA

We simulate 170 frequency maps between 115 and 199.5 MHz with spacings of 0.5 MHz. The maps consist of 512^2 pixels representing a comoving 1734 Mpc^2 area. At $z = 10.8$ this is equivalent to a field of view of $10^\circ \times 10^\circ$ or a resolution of $1.17 \text{ arcmin pixel}^{-1}$. Since an interferometer like LOFAR is insensitive to the mean value of the brightness temperature, we use mean-subtracted maps. We use the same set of simulations as Chapman et al. (2012) and the reader can refer there for more detailed information on the simulations.

The observable of the 21-cm radiation, the brightness temperature δT_b , is simulated using the seminumeric modelling tool 21CMFAST (Mesinger & Furlanetto 2007; Mesinger, Furlanetto

& Cen 2011). The code was run using a standard cosmology ($\Omega_\Lambda, \Omega_m, \Omega_b, n, \sigma_8, h$) = (0.72, 0.28, 0.046, 0.96, 0.82, 73) (Komatsu et al. 2011) and initialized at $z = 300$ on a 1800^3 grid. The velocity fields used to perturb the initial conditions as well as the resulting 21-cm δT_b boxes were formed on a cruder grid of 450^3 before being interpolated up to 512^3 . We define haloes contributing ionizing photons as having a minimum virial mass of $1 \times 10^9 M_\odot$.

The foreground simulations are obtained using the foreground models described in Jelić et al. (2008, 2010). We model Galactic diffuse synchrotron emission (GDSE), Galactic localized synchrotron emission, Galactic diffuse free-free emission and extragalactic foregrounds consisting of contributions from radio galaxies and radio clusters. While the Galactic foreground emission is simulated using Gaussian random fields, the total foreground signal is non-Gaussian. We do not consider the polarization of the foregrounds. The foregrounds simulated here can be up to five orders of magnitude larger than the signal we hope to detect but since interferometers such as LOFAR measure only fluctuations, foreground fluctuations dominate by ‘only’ three orders of magnitude.

To simulate the noise at each frequency, a LOFAR measurement set was filled with Gaussian noise in the uv plane. This was then imaged to create a real-space image, the root-mean-square (rms) of which can be normalized to the value as given by the prescription detailed in Chapman et al. (2012). For example the noise sensitivity at 150 MHz for an integration time of 600 h, a point spread function (PSF) of width 4 arcmin and a frequency spacing of 0.5 MHz is 64 mK. The 170 noise maps were uncorrelated over frequency, i.e. a different noise realization was used to fill the measurement set for each frequency.

The success of an interferometer such as LOFAR is highly dependent on how uv space is sampled. The particular pattern of uv sampling forms a beam which affects how the components such as the foregrounds are seen by the interferometer. Dirty foreground and 21-cm images were simulated by convolving with the PSF of the LOFAR set-up used to simulate the noise. The PSF used for creating dirty images (and for creating the noise as described in the previous section) was chosen to be the worst in the observation bandwidth, i.e. the PSF at 115 MHz. In observations the synthesized beam decreases in size with increasing frequency, causing point source signals to oscillate with the beam, producing a foreground signal with an oscillatory signal very much like that of the 21-cm signal (Vedantham, Shankar & Subrahmanyan 2012). However, this mode-mixing contribution has been found not to threaten the 21-cm recovery and have a power well below the 21-cm level (Bowman et al. 2006; Liu et al. 2009; Trott et al. 2012). As such we do not consider a frequency-dependent PSF here.

Once the foregrounds and 21-cm signal have been adjusted for uv sampling, the three component cubes are added together. The components of the total δT_b along a random line of sight are shown in Fig. 1.

4 RESULTS

In the following section, the word ‘reconstructed’ refers to a component which has been estimated from the simulated data using GMCA. The ‘residuals’ or (‘rest’ for short) are the difference between the total mixed signal and the reconstructed foregrounds and should therefore consist of the 21-cm signal, noise and any fitting errors.

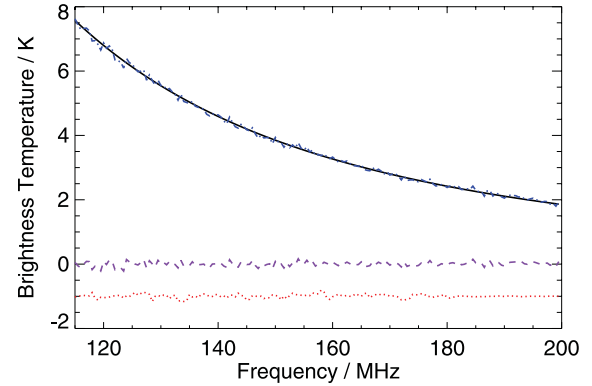


Figure 1. The redshift evolution of the simulated cosmological signal (red; dot), foregrounds (black; solid), noise (purple; dash) and total combined signal (blue; dash-dot). All components have undergone the PSF convolution. Note the 21-cm signal has been amplified by 10 and displaced by -1 K for clarity.

4.1 Source number

GMCA requires the specification of the number of sparse sources that the data can be defined by. We utilize this component separation technique to define the foregrounds in order to subtract them, treating the 21-cm signal as noise. Therefore, the number of sources refers to the number of foreground contributions which can be described by unique sparse descriptions (not necessarily the number of different foreground components such as Galactic free-free).

As each foreground model might be best described by a different number of sources, we seek to define the number of sources by minimizing the leakage of foregrounds into the 21-cm signal. Let us introduce the statistics \mathbf{L}_Y and \mathbf{R}_Y :

$$\mathbf{L}_Y = \mathbf{A}(\mathbf{A}^T \mathbf{A})^{-1} \mathbf{A}^T \mathbf{Y}, \quad (6)$$

$$\mathbf{R}_Y = \mathbf{Y} - \mathbf{A}(\mathbf{A}^T \mathbf{A})^{-1} \mathbf{A}^T \mathbf{Y}, \quad (7)$$

where \mathbf{A} is the mixing matrix calculated by GMCA and \mathbf{Y} is a data cube, for example the foregrounds or the 21-cm signal. \mathbf{L}_Y is the amount of the data \mathbf{Y} that contributes to the GMCA sources (in our case the reconstructed foregrounds). Thus we can calculate the amount of leakage of simulated noise and 21-cm signal, \mathbf{L}_{nocs} into the reconstructed foregrounds by allowing \mathbf{Y} to equal the combined simulated 21-cm and noise data cube. Conversely, \mathbf{R}_Y is the amount of data \mathbf{Y} which does not contribute to the GMCA source model. For example, letting \mathbf{Y} equal the simulated foreground cube, \mathbf{R}_{fg} , will tell us how much of the foregrounds leak into the residuals.

We take the power spectra of \mathbf{L}_{nocs} and \mathbf{R}_{fg} and compare them to the power spectra of the 21-cm signal to see the number of sources for which leakage is minimized (Fig. 2). Note that, prior to our wavelet-type analysis in Section 4.2, we adopt the default setting of GMCA for this source number analysis, the IUWT using the à trous algorithm.

To be confident of our reconstructed 21-cm signal, we ideally want both the power spectra of \mathbf{R}_{fg} and \mathbf{L}_{nocs} to lie below that of the 21-cm signal and to be as small as possible. We see that, while one source does not seem enough to accurately constrain the foregrounds, there is very little difference between the leakages resulting from a two or three source foreground model. Indeed, we find this holds true for four and five sources also. As more and more sources are added to the model, we might expect the 21-cm signal itself to leak into the foreground model and be picked out as an

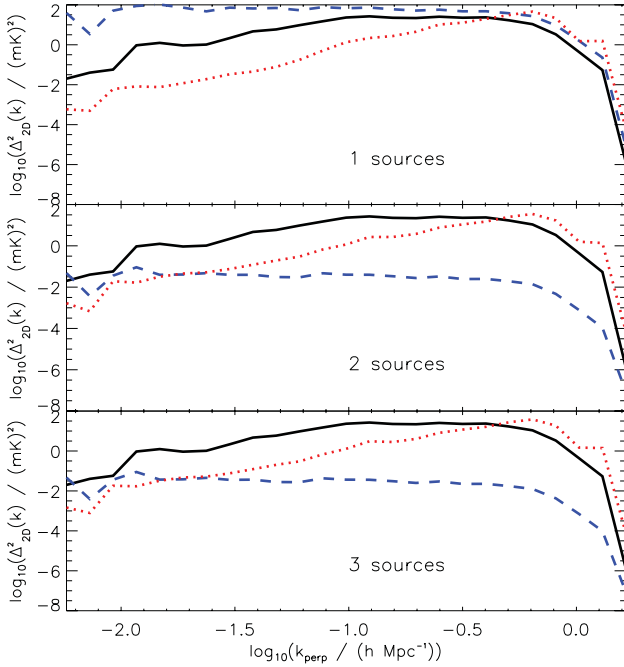


Figure 2. The 2D power spectrum of \mathbf{R}_{fg} (blue dash) and \mathbf{L}_{nocs} (red dot) and the 21-cm power spectrum (black, solid) at a frequency of 160.0 MHz for GMCA with one, two and three sources (top, middle and bottom, respectively).

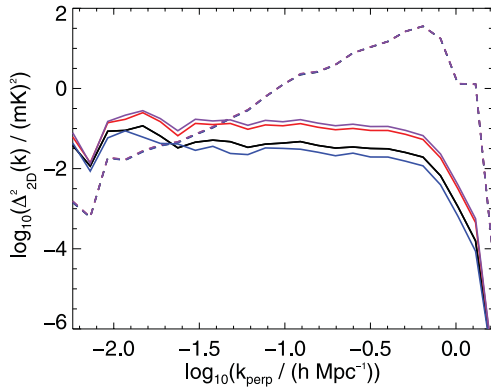


Figure 3. The 2D power spectrum of \mathbf{R}_{fg} (solid) and \mathbf{L}_{nocs} (dashed) at 160 MHz for IUWT (black), Mallat's wavelet transform (red), Feauveau's wavelet transform without undersampling (blue) and Haar's wavelet transform (purple). Note that the dashed lines all lie on top of each other.

individual source. However the magnitude of this (seen as part of \mathbf{L}_{nocs}) will be small as the signal-to-noise ratio of the foregrounds is much larger than that of the cosmological signal. When applying to real data we will have to rely on models of the foregrounds as observed by the data to estimate the level of leakage from the signal on to the foregrounds.

We choose to assume two foreground sources for the rest of this paper, though the reader should be aware that different foreground models may require different source numbers in order to optimize the method.

4.2 Choice of wavelet

In Fig. 3 we consider how the leakages \mathbf{R}_{fg} and \mathbf{L}_{nocs} change depending on the choice of wavelet used by GMCA. There are many different

types of wavelets, but they can be broadly categorized by whether they are decimated (i.e. provide a redundant signal representation), isotropic and which filter they use for the separation of data at different scales. Here we consider the IUWT, a decimated wavelet transform (referred to as Mallat), a non-dyadic and undecimated wavelet transform (referred to as Feauveau) and an undecimated wavelet transform using the Haar filter (referred to as Haar).

We can see that the choice of wavelet can affect how small the foreground leakages are, and therefore the success of the method. These differences are highly dependent on frequency as well, with one wavelet type out-performing others at certain frequencies. For our implementation we choose the IUWT as it consistently minimizes the leakages over the frequency range. We will show in Section 4.3.2 that we can potentially correct for \mathbf{L}_{no} , though not \mathbf{L}_{cs} and \mathbf{R}_{fg} due to the blind nature of the problem.

4.3 Power spectra

EoR experiments aim to recover the power spectrum of the cosmological signal over a broad range of frequencies.

The power spectrum of a line-of-sight/map/cube at a single frequency is calculated by 1D/2D/3D Fourier transforming that line-of-sight/map/cube and binning the pixels according to Fourier scale, k . The power at any particular k , $\langle \delta(k) \delta^*(k) \rangle$ is the average power of all the uv cells in the bin centring on k . The error on the point for a particular bin, k_i , is calculated as $\sigma_i = \frac{\langle \delta(k_i) \delta^*(k_i) \rangle}{\sqrt{n_{k_i}}}$, where n_{k_i} is the number of uv cells that reside in that k bin. The power spectrum of the reconstructed 21-cm signal is calculated by subtraction of the noise power spectrum from the GMCA residuals power spectrum. The total error on the reconstructed 21-cm power spectrum is calculated using the above error formula applied to the reconstructed 21-cm power spectrum, added in quadrature with the formula applied to the noise power spectrum, in order to take into account both sample variance and the effect of any error in the noise estimate. Note that we assume Gaussianity whereas the 21-cm signal is not Gaussian and also we calculate the error bars from the power of a single realization rather than over an ensemble of simulations. We ask the reader to bear in mind that these error bars might be considered incomplete because of this.

To explain a few graphical conventions: any points where the power of the residuals is below the power of the noise are omitted, as this leads to an unrealistic negative reconstructed 21-cm power; any error bars extending to below the x -axis in linear space are shown with a lower error bar of equal length to the upper error bar in log space.

4.3.1 1D power spectra

The 1D power spectra are calculated over frequency wedges of 8 MHz to avoid evolution effects. Each line of sight produces a 1D power spectrum, one for each of the 512×512 pixels. These power spectra are then averaged over all the pixels. The frequencies mentioned correspond to the frequency in the middle of each 8-MHz wedge and the quantity plotted in Fig. 4 is $\Delta_{\text{1D}}^2(k) = \frac{Lk \langle \delta(k) \delta^*(k) \rangle}{\pi}$, where L is the comoving length of the wedge. The 1D power spectra are recovered to high accuracy across the frequency range and across the scales.

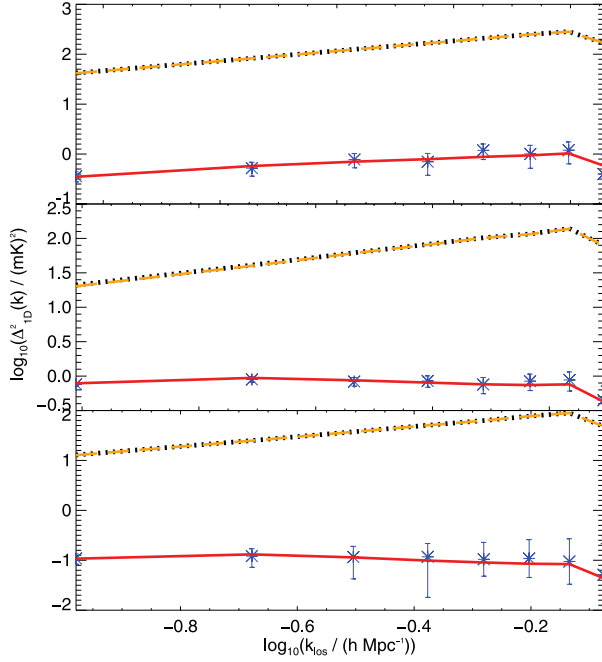


Figure 4. The 1D power spectra of the simulated 21-cm (red, solid), the residuals (black, dot), the reconstructed 21-cm (blue, points) and the noise (orange, long dash). Three 8 MHz frequency wedges centred at 127, 151 and 175 MHz, respectively, are shown from top to bottom.

4.3.2 2D and 3D power spectra

For the 2D power spectra, the quantity plotted is $\Delta_{2D}^2(k) = \frac{Ak^2(\delta(k)\delta^*(k))}{2\pi}$, where A is the area of the simulation map. To calculate the 3D power spectra we divide the cube into subbands of 8 MHz to avoid signal evolution effects. The quantity plotted is $\Delta_{3D}^2(k) = \frac{Vk^3(\delta(k)\delta^*(k))}{2\pi^2}$, where V is the volume of the subband.

We now consider the recovered 2D and 3D 21-cm power spectra and how best we can minimize the effects of leakage. The noise leakage into the reconstructed foregrounds can be accurately quantified by creating an independent realization of the noise (no2) (for real data it is assumed we will know the statistics of the noise to high precision and can therefore create a second realization from the known noise power spectrum) and applying equation (6) to find \mathbf{L}_{no2} . We find that the power spectra of \mathbf{L}_{no} and \mathbf{L}_{no2} are almost identical meaning that \mathbf{A} is not a strong function of the original noise realization. We can use this information to find a better estimate of the 2D and 3D power spectra and compensate for the noise leakage using the following calculation:

$$\tilde{\Delta}_{csrec}^2 = \Delta_{rest}^2 - \Delta_{no}^2 + \Delta_{Lno2}^2. \quad (8)$$

To calculate a leakage ratio we would generally have to include all the leakages, such that leakage ratio = $\frac{\Delta_{cs}^2}{\Delta_{Rfg}^2 + \Delta_{Lnoc}^2 + \Delta_{cs}^2}$. However, since we can correct for \mathbf{L}_{no} we can instead calculate

$$\text{leakage ratio} = \frac{\Delta_{cs}^2}{\Delta_{Rfg}^2 + \Delta_{Lcs}^2 + \Delta_{cs}^2}, \quad (9)$$

which quantifies the amount of 21-cm leakage into the foregrounds and vice versa. We can see the leakage ratio plotted for the 2D and 3D power spectra for multiple frequencies in Fig. 5. We see that we can expect an accurate 21-cm power spectrum recovery at large k

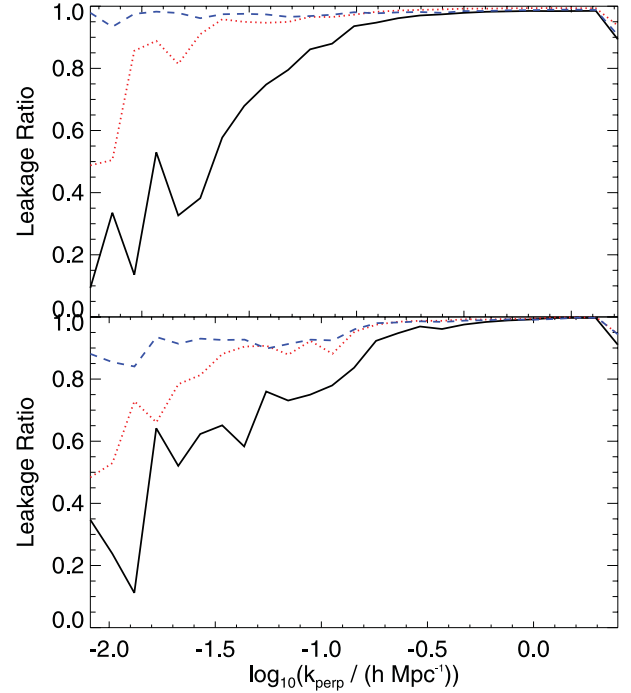


Figure 5. Top: the leakage ratio for the 2D power spectra for frequencies of 130 MHz (black, solid), 150 MHz (red, dot) and 170 MHz (blue, dash). Bottom: the leakage ratio for the 3D power spectra for frequencies of 135 MHz (black, solid), 151 MHz (red, dot) and 167 MHz (blue, dash).

once the noise leakage is compensated for. The smaller k recovery is still very dependent on the foreground magnitude and therefore frequency of observation.

We can see the result of applying equation (8) in Figs 6 and 7. Wherever \mathbf{R}_{fg} and/or \mathbf{L}_{noc} exceeds the power of the simulated 21-cm signal we see a degraded fit in Fig. 6. The 2D power spectra are recovered to excellent accuracy and we see that once the noise leakage is taken into account there is much less leakage at large k scales, allowing a more complete power spectrum reconstruction.

For the 3D power spectra, a similarly accurate recovery is made across the frequency range. The recovery in 3D is more precise due to the larger amount of data in a box as opposed to a single slice. Again, once the noise leakage is taken into account the recovered spectra become much more complete on the larger k scales.

5 PHASE CONSERVATION AND IMAGING

Recently it has been shown that imaging of the neutral hydrogen in the late stages of reionization is possible with the current generation of radio telescopes when angular scales larger than 0.5° are considered, independent of the type of reionization source (Zaroubi et al. 2012). Here, we compare the output residual maps with the simulated 21-cm maps and consider how well the phases of the 21-cm signal are conserved through the foreground removal process. The better the phases are conserved, the more correlation between maps we will observe. We will also consider the maps at different scales and as such we also decompose the output maps into eight wavelet scales using the IUWT.

For a particular frequency, we calculate the phase of each uv point in a Fourier transformed map, F , as $\text{Phase}[u, v] = \tan^{-1}(\text{Im}(F(u, v))/\text{Re}(F(u, v)))$.

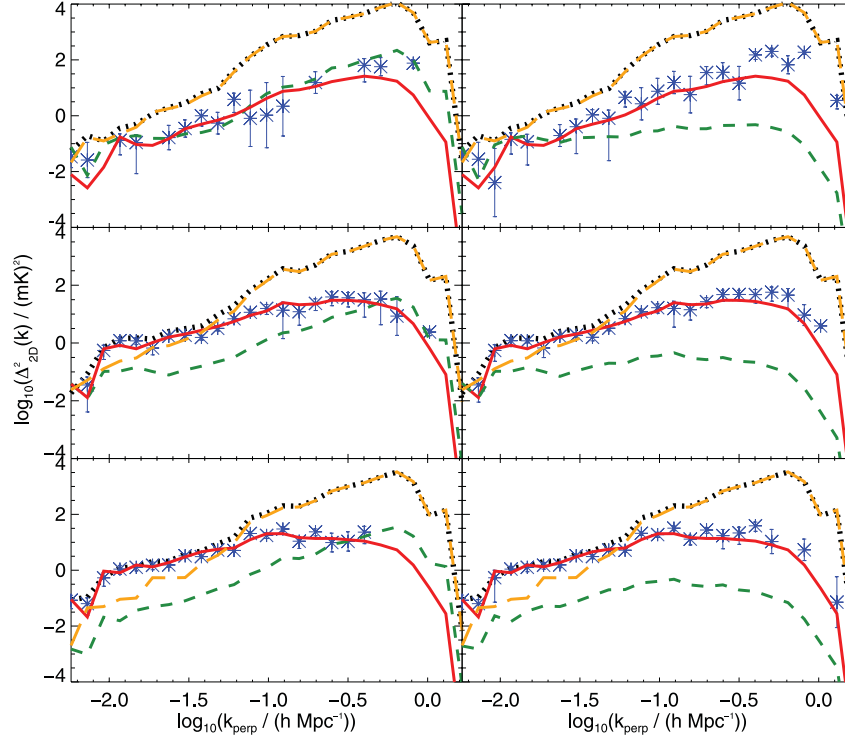


Figure 6. 2D power spectrum of the simulated 21-cm signal, reconstructed 21-cm signal, residuals and noise at 130 MHz, or $z = 9.92$, 150 MHz, or $z = 8.47$, and 170 MHz, or $z = 7.35$, from top to bottom. The left-hand column is the fiducial data whereas the right-hand column plots the reconstructed 21-cm power spectrum but with the leakage determined from the second noise realization added, as described in Section 4.3.2. Line styles are as described in Fig. 4 with the additional dark green dashed line representing the total leakage power ($\Delta_{R_{\text{fig}}}^2 + \Delta_{L_{\text{nos}}}^2$) in the left-hand column and the leakage assuming noise leakage has been corrected ($\Delta_{R_{\text{fig}}}^2 + \Delta_{L_{\text{cs}}}^2$) in the right-hand column.

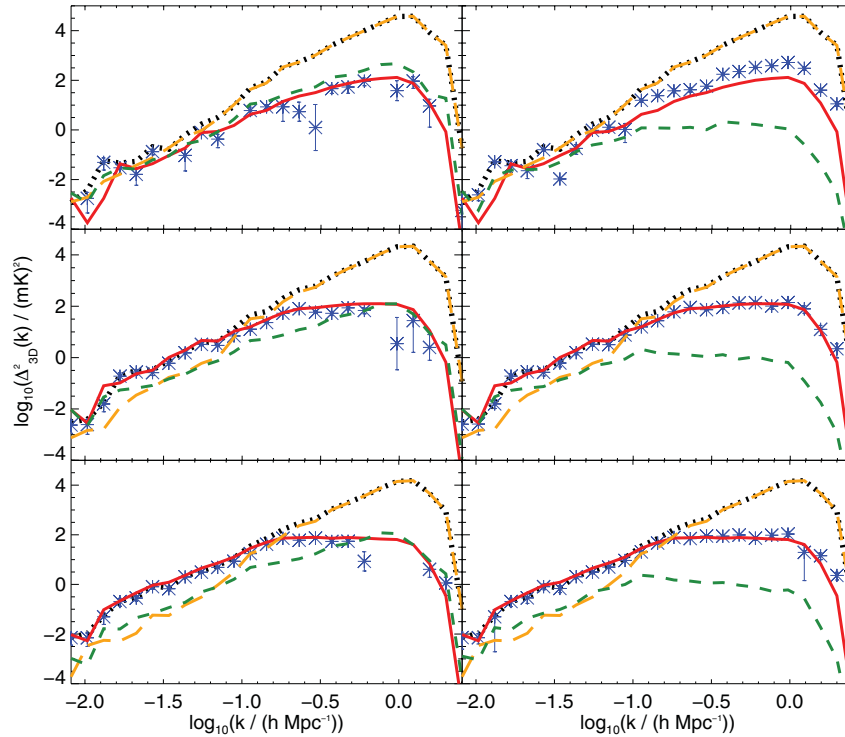


Figure 7. 3D power spectrum of the simulated 21-cm signal, reconstructed 21-cm signal, residuals and noise at 135 MHz, or $z = 9.51$, 151 MHz, or $z = 8.40$, and 167 MHz, or $z = 7.50$, over an 8-MHz subband (top to bottom). The left-hand column is the fiducial data whereas the right-hand column plots the reconstructed 21-cm power spectrum but with the second noise realization leakage added. Line styles are as described in Fig. 6.

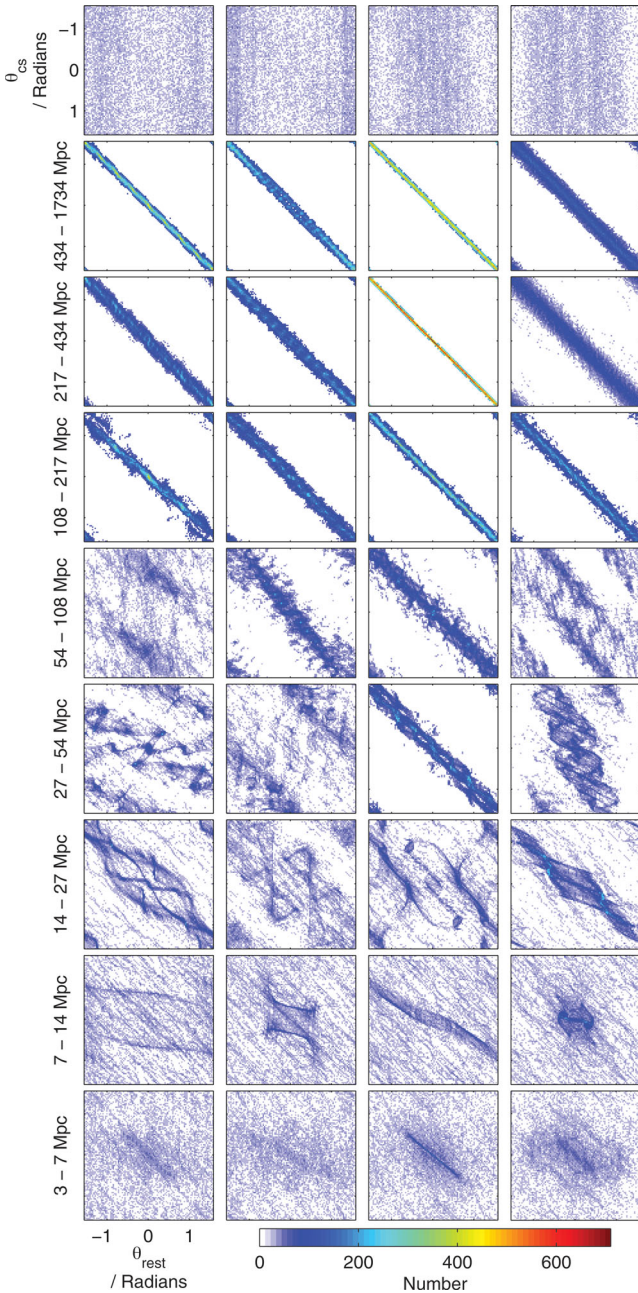


Figure 8. Density maps of the phase of maps of the simulated 21-cm and the residuals. From left to right are maps at frequencies 130, 145, 160 and 175 MHz. From top to bottom are maps of the complete cubes and then of increasingly small-scale wavelet scales. A clear diagonal signifies excellent phase recovery and therefore clearer images can be recovered. We see that on scales above 108 Mpc, the phases are well preserved; on smaller scales, however, the phases are highly uncorrelated. It is clear that considering different wavelet scales can result in much better phase recovery than considering the full cube.

In Fig. 8 we see the phase density relating to each wavelet scale of the 21-cm and residual cubes. For each frequency we calculate the phase of each pixel in the 21-cm and residual maps and use this as a coordinate on a phase density map where the x -axis is the phase of the residuals and the y -axis is the phase of the 21-cm map. The more pixels with coordinates corresponding to a particular bin in the

phases, the higher the phase density we will observe. If GMCA perfectly preserves the phase of the 21-cm signal we should see a diagonal phase density plot. For clarity, only the phase bins with a pixel count in the largest 67 percent of the pixel count distribution are plotted. We see that for the three crudest wavelet scales there is excellent phase recovery across the frequency range, while at 160 MHz this excellent recovery can be seen on even smaller wavelet scales. This is in line with expectations since the relative rms values (i.e. ratio of the rms of the 21-cm and noise and ratio of the rms of the 21-cm and foregrounds) for the 21-cm signal both peak at just after 160 MHz.

In Fig. 9 we start with the crudest wavelet scale and then add the next crudest scale one at a time to see the effect on phase conservation, finding that this improves the recovery considerably.

In Fig. 10 we compare the reconstructed signal maps with the simulated signal maps at the different wavelet scales. It is clear that the foregrounds are reconstructed to a high accuracy at all scales. The residuals show clear correlation with the 21-cm maps – especially at scales 54–434 Mpc. By considering the data on different scales we compensate for the small-scale noise leakage and can retrieve convincing reconstructed images in comparison to the map with all scale data included.

We plot the Pearson correlation coefficient between the simulated 21-cm maps and the residual maps at different individual scales in Fig. 11. On an individual basis, we can clearly see that the finer the scale of structure we correlate, the less of a correlation there is between the residuals and simulated 21-cm maps. The finer wavelet scale we look at, the more dominant noise leakage will be in the residual maps and so a smaller correlation is observed. By only comparing the crudest wavelet scales however, we risk losing a lot of the small-scale 21-cm structure and being increasingly dominated by the foreground signal.

We can add several of the wavelet scales together in order to balance having enough useful information without including too much noise leakage on the smaller wavelet scales. To compare with Zaroubi et al. (2012), we also include images of the full data which have been smoothed with a 20-arcmin Gaussian kernel. Wavelet decomposition has the advantage of providing a selection of scales on which one can analyse the images, as opposed to being restricted by filters such as a Gaussian kernel which simply remove all modes below a certain scale. However, the scales at which one can analyse images with wavelet decomposition are determined by the method itself – you cannot then ask what the data look like at a scale half way in-between two wavelet scales.

In Fig. 12 we recover impressive images of the reionization signal at 165 MHz when the smallest scale information is discarded. Comparing the residual and 21-cm maps on each row we find correlation coefficients of 0.689, 0.687 and 0.588 for the top, middle and bottom rows, respectively. We therefore conclude that the wavelet decomposition more optimally removes the noise from the residuals than the smoothing technique (bottom row) employed by Zaroubi et al. (2012). In Fig. 10 we can see the wavelet decomposition has the effect of clustering the noise around the edges of the image (this effect is particularly pronounced in rows 3 and 4). To avoid this edge effect we also correlate the maps in Fig. 12 again but this time considering only the pixels in a central patch covering 50 percent of the total map area. We find correlation coefficients of 0.905, 0.788 and 0.605 for the top, middle and bottom rows respectively. With real data, we can assume that this edge effect would not be a problem as the fitting and decomposition can always be carried out over a slightly larger cube with only the central region being used for data analysis.

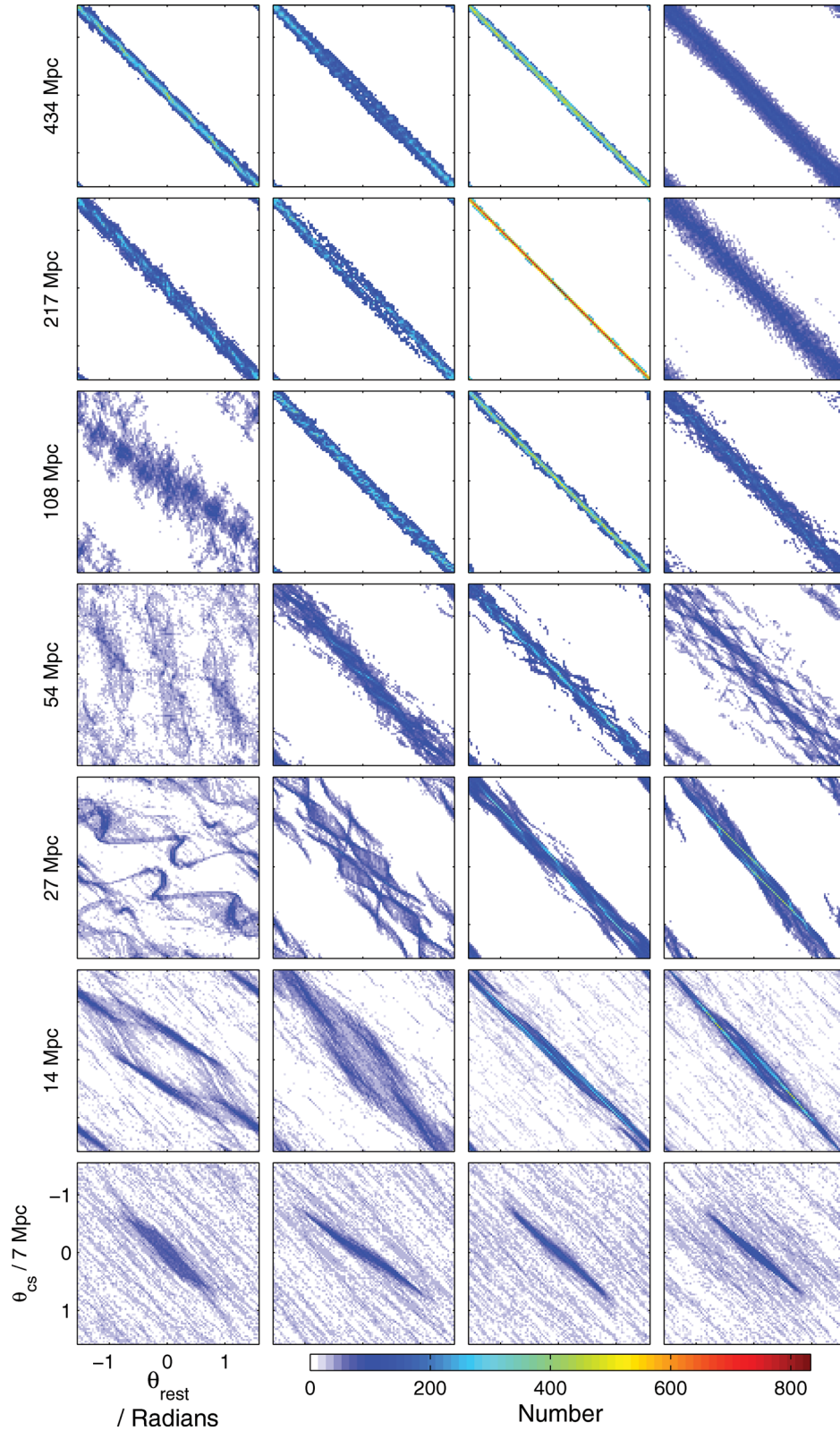


Figure 9. Density maps of the phase of maps of the simulated 21-cm and the residuals. From left to right are maps at frequencies 130, 145, 160 and 175 MHz. From top to bottom are maps of cubes with only the crudest scale present and then of only the 2, 3, 4, 5, 6 and 7 crudest wavelet scales. The minimum distance scale information included is labelled for each wavelet scale, the maximum is always 1734 Mpc. A clear diagonal signifies excellent phase recovery and therefore clearer images can be recovered. The addition of several scales together results in clearer diagonals than considering scales individually in Fig. 8.

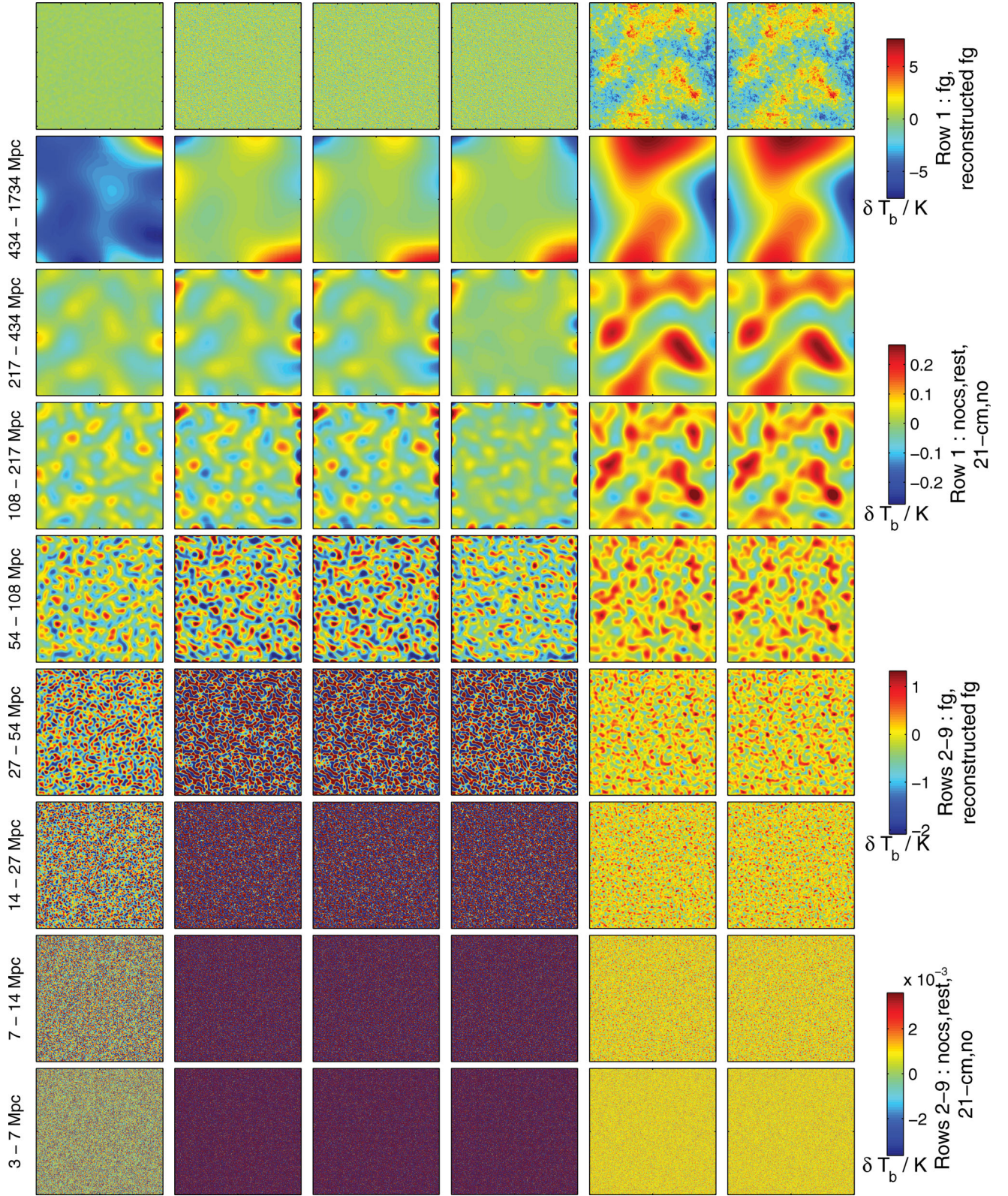


Figure 10. The decomposition of the 21-cm signal, residuals, simulated noise + 21-cm signal, noise, foregrounds, reconstructed foregrounds and noise (left to right). From top to bottom, the rows are the original image at 165 MHz, and then the wavelet decomposition of this image at the eight wavelet scales. We can see that the simulated and reconstructed foregrounds have a high correlation at all scales and even in the full cube. Similarly, the noise + 21-cm and residuals also share this strong correlation. As we cannot remove the noise directly we must look for a correlation between the residuals and simulated 21-cm, which will come as a result of little or no correlation between the noise and residuals at certain scales. The noise dominates too much in the full cube and on the large k scales, however, we can clearly see a correlation by eye on distance scales between 108 and 434 Mpc. At the largest scale, the 21-cm signal is so small that the residuals are dominated by noise.

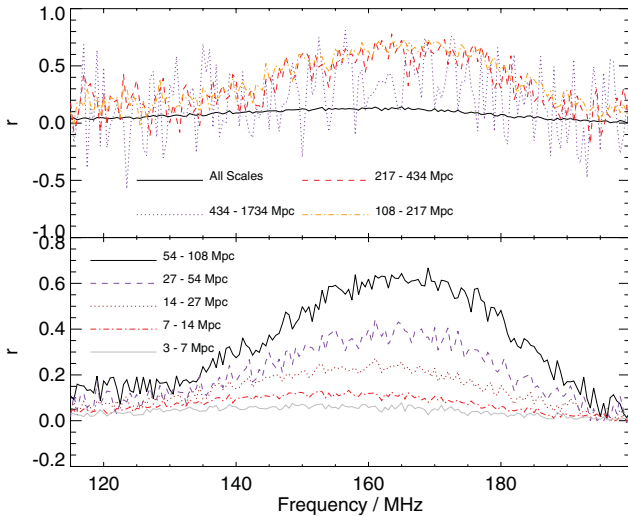


Figure 11. The Pearson correlation coefficient between 21-cm and residual maps. There is very little correlation when all scales are present or on the smallest scales, however, we reach correlation coefficients of over 0.6 for distance scales 54–434 Mpc. The correlations are always much weaker at the lower end of the frequency range because the noise and foregrounds are at their highest and at the higher end of the frequency range because the 21-cm signal is negligible.

6 CONCLUSIONS

In this paper we have assessed the sparsity-based blind source separation technique GMCA as a possible way of removing the foregrounds on a 21-cm EoR signal. We recover the 1D, 2D and 3D power spectra to high accuracy across the frequency range. Since the mixing matrix calculated by GMCA was shown not to be a strong function of the noise realization, we were able to compensate for leakage of noise power into the reconstructed foregrounds using an independent noise realization, leading to more complete 2D and 3D power spectra.

We also considered if images of reionization could be recovered from the LOFAR-EoR data once foreground removal with GMCA has been carried out. Using wavelet decomposition, we considered the phase correlation between the GMCA residuals and the simulated 21-cm at different scales. We find strong correlations at the cruder wavelet scales and add several scales together to balance the amount of information in an image with the accuracy of the phase recovery. We find that when distance scales of below 54 Mpc are discounted, the GMCA residuals images are highly correlated with the 21-cm images, with correlation coefficients of just less than 0.7. In comparison, smoothing the images did not produce as strong a correlation.

GMCA is a highly adaptable method and there remains the possibility that with careful tuning, the 21-cm signal could be picked out as a separate source as opposed to being present as a residual of the process. We intend to explore this further and consider using different mixing matrices for each scale.

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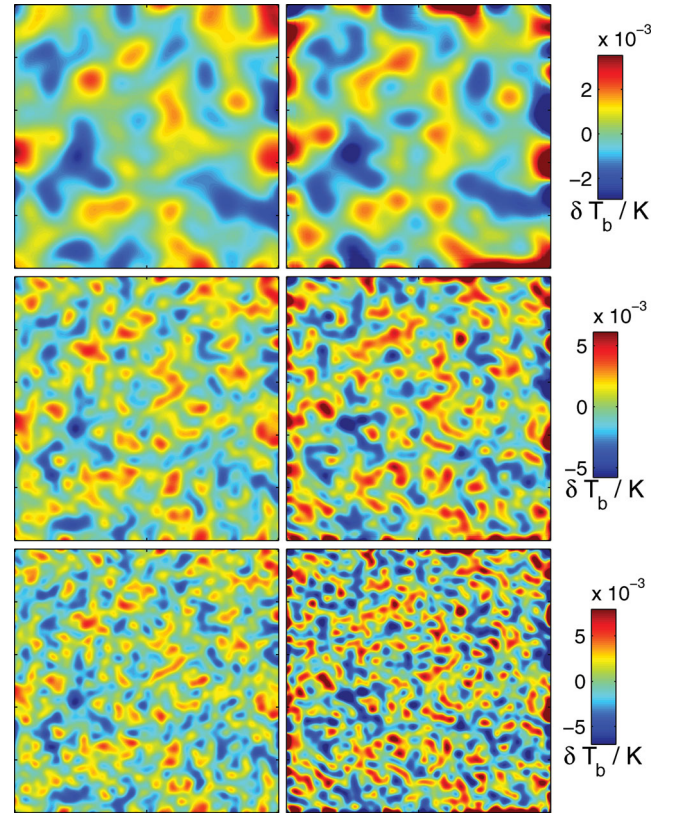


Figure 12. In the left-hand column we show the 21-cm signal and in the right-hand column the residuals of GMCA at 165 MHz. In the top row only distance scales between 1734 and 108 Mpc are included, in the middle only distance scales between 1734 and 54 Mpc are included and on the bottom the images with all scales present have been smoothed with a 57 Mpc (≈ 20 arcmin) Gaussian kernel. Clear correlations can be seen between the columns (coefficients of 0.689, 0.687 and 0.588 for the top, middle and bottom rows, respectively). Considering only the pixels in a central patch covering 50 per cent of the total map area we find correlation coefficients of 0.905, 0.788 and 0.605 for the top, middle and bottom rows, respectively.

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